



Article

Numerical Analysis of CNC Milling Chatter Using Embedded Miniature MEMS Microphone Array System

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Abstract: With the increasingly common use of industrial automation for mass production, there are many computer numerical control (CNC) machine tools that require the collection of data from intelligent sensors in order to analyze their processing quality. In general, for high speed rotating machines, an accelerometer can be attached on the spindle to collect the data from the detected vibration of the CNC. However, due to their cost, accelerometers have not been widely adopted for use with typical CNC machine tools. This study sought to develop an embedded miniature MEMS microphone array system (Radius 5.25 cm, 8 channels) to discover the vibration source of the CNC from spatial phase array processing. The proposed method utilizes voice activity detection (VAD) to distinguish between the presence and absence of abnormal noise in the pre-stage, and utilizes the traditional direction of arrival method (DOA) via multiple signal classification (MUSIC) to isolate the spatial orientation of the noise source in post-processing. In the numerical simulation, the non-interfering noise source location is calibrated in the anechoic chamber, and is tested with real milling processing in the milling machine. As this results in a high background noise level, the vibration sound source is more accurate in the presented energy gradation graphs as compared to the traditional MUSIC method.

Keywords: beamforming; direction of arrival; voice activity detection; chatter detection

1. Introduction

In recent years, industrial exhibitions have presented the latest measuring instruments and systems for Industry 4.0, including those in the area of electronic acoustics. Due to their wide variety of frequency responses and their use in huge environmental spaces, microphone arrays have continuously been applied in new ways, allowing for novel uses of a relatively old technology. In the past, the central processing unit (CPU) speed of electronic acoustic digital sensors was slow and inaccurate, while analog sensors, which required large and bulky equipment, were expensive. With the semiconductor industry booming in recent years, however, CPU processing has become cheaper and faster, allowing digital sensors to be modularized easily. Step by step, the application of digital signal processing (DSP) has allowed various hardware restrictions to be overcome, making it possible to equip digital sensors with various advantages, such as microminiaturization, heavy computing power, precision, and high processing speed. DSP can be applied on the industry automation production line.

The improvement of computer numerical control (CNCs) has recently focused on providing automated and user-friendly measurements, which requires the use of a sensor or sensors to collect data regarding unstable conditions when cutting. During the operation of intelligent CNCs, lots of unsteady flutters appear, and these flutters result from the vibration of the end mill. If the end mill

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passes through the cut surface again, the cut will go deeper and deeper, and eventually, the CNC will break down. In this way, bad milling processing will result. In order to solve this processing problem, an accelerometer can be attached to the spindle to collect data regarding the detected vibration of the CNC. However, chatter usually occurs between the end mill and the workpiece, and it is difficult to attach an accelerometer on the milling surface. Simultaneously, the noise will combine with the milling chatter [1–3]. In light of these issues, in this study, we used an embedded MEMS microphone array system as a means of applying acoustic localization estimation to intelligent CNCs [4].

Early research into the technique of using sensor arrays to identify the direction of arrival (DOA) [5] involved the use of such arrays in high-frequency radar systems. This approach can estimate the direction of a source according to the power and phase of the signal. Recently, this technique has usually been used with microphone arrays to estimate the audible band of a sound source. There are three kinds of estimation of source. The first kind usually uses the DOA to locate the source direction. In this approach, the specific direction energy is strengthened by the beamforming of the DOA method in the time domain, which can be accomplished using the steer beamformer and delay-and-sum (DAS) methods. The second kind is the time difference of arrival (TDOA) approach [6]. Usually, this approach involves the combined use of a couple of microphones and utilizes the direction and angle relation between the source and the microphones to estimate the source's location. The third kind consists of using the least-squares (LS) method and cross correlation (CC) to calculate the relation between two signals from a microphone. This approach must be used in a steady environment and must be conducted via digital signal processing in a noisy environment. Therefore, it is proposed that the generalized cross correlation (GCC) [7] and the time-domain phase transform generalized cross correlation (GCC-PHAT) [8] be used in this complex situation.

The proposed method consists of processing the orthogonal relationship of eigenvectors from multiple signals in order to estimate the direction of the source. The current study also used the multiple signal classification (MUSIC) algorithm [9–11] to estimate the sound source. The precision of direction estimation when using the eigenstructure method will be affected by the noise of the surrounding environment. Previous studies [12–15] have shown that voice activity detection (VAD) has already been used for many years to implement noise cancellation for voice signals. In this study, we propose the voice activity detection (VAD) method to subtract the non-sound source signal energy and eliminate the unsteady noise in the spectrum. After the VAD process, the MUSIC algorithm is used to estimate the sound location. The simulation results of this study are based on the CNC milling chatter signal in a high frequency region with simulated noise. The results are validated with numerical simulation and tested by carrying out real milling processing in the CNC machine.

2. Method Description

VAD has already been used for many years to implement noise cancellation for voice signals, and it is also highly suitable for one-dimensional blind signals that contain noise elements. As described in the introduction section, the objective of this study was to use VAD to separate noise and target signals and eliminate the noise in the spectrum, after which the pure signals were input into the sound source location to carry out calculations using an estimation algorithm, thereby identifying the effective sound source location. The primary method utilized in this study was the robust VAD method proposed by Sohn [15]. The method can be expressed using the equation

$$H(e^{j\omega}) = X(e^{j\omega}) + N(e^{j\omega})$$
(1)

where $e^{j\omega}$ is the frequency domain signal obtained through the Fourier transform. $X = [x_1, x_2, ..., x_m]$ is the clean signal vector obtained from the m microphones, $N = [n_1, n_2, ..., n_m]$ is the vector

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corresponding to the noise detected by each microphone. Using a Gaussian statistical model, the following two hypothesis functions are obtained

$$P(H(e^{j\omega})|N(e^{j\omega})) = \frac{1}{\lambda_{N,e^{j\omega}}} e^{\left(-\frac{|H(e^{j\omega})|^2}{\lambda_{N,e^{j\omega}}}\right)}$$
(2)

and

$$P(H(e^{j\omega})|[X(e^{j\omega}) + N(e^{j\omega})]) = \frac{1}{\lambda_{N,e^{j\omega}} + \lambda_{X,e^{j\omega}}} e^{\left(-\frac{|H(e^{j\omega})|^2}{\lambda_{N,e^{j\omega}} + \lambda_{X,e^{j\omega}}}\right)}$$
(3)

 $\lambda_{N,e^{j\omega}}$ and $\lambda_{X,e^{j\omega}}$ are the variances for the target signal and noise signal. By rearranging Equations (1)–(3), the likelihood coefficient at the frequency point can be expressed as

$$\Lambda(e^{j\omega}) = \frac{P(H(e^{j\omega})|N(e^{j\omega}))}{P(H(e^{j\omega})|[X(e^{j\omega}) + N(e^{j\omega})])} = \frac{1}{1+\alpha}e^{\frac{\alpha \cdot \beta}{1+\alpha}}$$
(4)

in which $\alpha=\lambda_{\mathrm{X},e^{j\omega}}/\lambda_{N,e^{j\omega}}$ and $\beta=H(e^{j\omega})/\lambda_{N,e^{j\omega}}$ are the signal-to-noise ratio (SNR) coefficients of the a priori and a posteriori signals, respectively. Therefore, after obtaining the logarithm of $\Lambda(e^{j\omega})$, an attempt is made to maximize the SNR coefficient of the priori signal α , and the Itakura–Saito method can be used to derive the objective function as

$$J = \log \Lambda(e^{j\omega}) = \frac{1}{m} \sum_{i=0}^{m-1} (\beta - \alpha) = \frac{1}{m} \sum_{i=0}^{m-1} (\beta - \log(\beta) - 1)$$
 (5)

then, using the Newton's method iteration, we obtain the equation

$$\beta(k) = a \frac{\Lambda(e^{j\omega})[k-1]}{\lambda_N(k-1)} + (1-a) \cdot (\beta[k-1]-1)$$
(6)

where $k(=0,1,2,...,k_n)$ is the iteration number and a is a small iteration step (a << 1). By using the least square method or repeated iteration to solve Equation (6), Equation (5) can be minimized to allow for α and β to be substituted into Equations (2) and (3), through which the target signal and noise signal values can be obtained. By applying the inverse Fourier transform, these values can be converted to the time domain signal X(t). Then, the MUSIC algorithm is applied to obtain the estimated value of DOA [16] from the following estimation of the covariance matrix R(X),

$$R(X) = \frac{1}{M} \sum X \cdot X^{H} = \frac{1}{M} \sum BS \cdot (BS)^{H}$$
 (7)

where $S = [s_1, s_2, ..., s_m]$ is the strength of the microphone signal, $B = [b(\theta_1, \varphi_1), b(\theta_2, \varphi_2), ..., b(\theta_m, \varphi_m)]$ The next step is to carry out eigenvalue decomposition (EVD), and the relevant equation can be written as

$$R(X) = B \cdot E\{S \cdot S^H\} \cdot B^H + \sigma^2 I \tag{8}$$

The spherical coordinates of the sound source locations can be defined by the spatial spectrum $P(\theta, \phi)$ as

$$P(\theta, \varphi) = \frac{1}{b^H(\theta_i, \varphi_i)VV^Hb(\theta_i, \varphi_i)} = \frac{1}{\|V^HB\|^2}$$
(9)

when matrix B is orthogonal with respect to each column of V, the eigenvalue and eigenvector can be solved by $B^HV_{\delta}=0$. If $B^HV_{\delta}\neq 0$, the eigenvector indicates the presence of existence noise, and the denominator of Equation (9) is not zero. The peak value of $P(\theta, \varphi)$ can be determined by changing the values of (θ, φ) of Equation (9).

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3. Results and Discussion

3.1. System Setup

This study proposed an embedded system architecture consisting of three parts. First, the microphone array system was created and fixed in a stable manner to the hardware. The signal was then recorded by beamforming with an embedded system, as shown in Figure 1. Second, file transfer protocol (FTP) wireless transmission was used to transfer the data to a personal computer (PC) for analysis. Third, after the noise cancelling calculation and source estimation were conducted in the post-processing, the signal was then shown in the presented energy gradation graphs. The signals of the microphone array were collected by Xilinx Field-Programmable Gate Array (FPGA), which allowed the data from the array to approach synchronous transmission. The structure of proposed method is shown in the Figure 2.

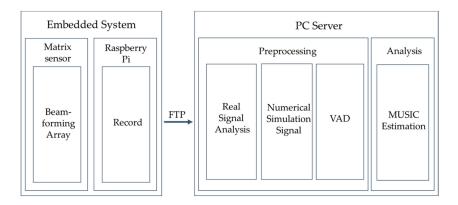


Figure 1. The measurement system setup and numerical analysis process.

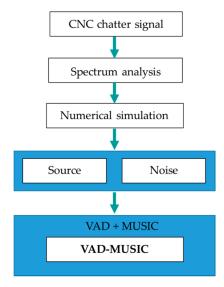


Figure 2. The structure of the proposed method for noise reduction of source direction estimation.

3.2. Comparison Results

In the first test, we use the MUSIC algorithm to estimate the different azimuths and elevation angles of the CNC milling sound source. The three different spherical coordinates of the 1 kHz sound source locations are defined as $(\theta, \varphi) = (-100, 0)$, $(\theta, \varphi) = (0, 30)$, and $(\theta, \varphi) = (100, 60)$. The results for the sound sources are shown in Figure 3a–c. It is observed that the results show the accurate location of sound sources as a peak in each figure.

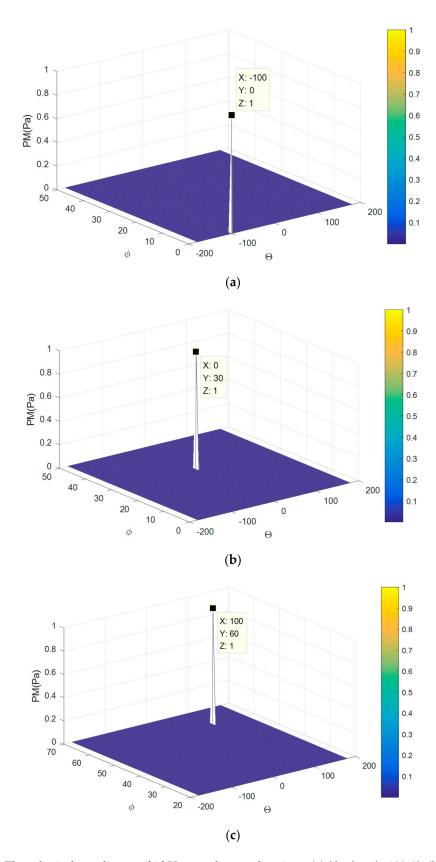


Figure 3. The spherical coordinates of 1 kHz sound source locations: (a) $(\theta, \varphi) = (-100, 0)$; (b) $(\theta, \varphi) = (0, 30)$; (c) $(\theta, \varphi) = (100, 60)$.

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For the next validation case, the source estimation of the circumstances of the CNC milling chatter concerns multi-tone simulation. The settings of the CNC milling operation are shown in Table 1, including four different combinations of spindle speed, depth of cut and feed rate. The related results of the spectrum analysis are shown in Figure 4a–d. The results of the resonance frequencies indicate that the different circumstances of CNC milling chatter are observed in the high frequency range of 3.4–4.5 kHz. The results in Figure 4a–d show that the peak intervals of the resonance frequency are almost consistently separated by 134 Hz.

Based on the frequency domain results in Figure 4d, we use the time domain signal with additive noise from different SNRs as 12, 6, and 0 dB (as shown in Figure 5a–c), so as to evaluate the performance of CNC milling sound source estimation as $(\theta, \varphi) = (-100, 30)$ and to compare the original MUSIC and VAD-MUSIC approaches. As indicated in Figure 6a–c, the original MUSIC approach yields unstable results as the energy of the white noise is increased. After the energy level was increased, the sound source locations obtained by the highest peak for SNR = 12, 6, and 0 dB corresponded to $(\theta, \varphi) = (-96, 36), (-99, 47)$, and (-99, 65) in Figure 6a–c, respectively. It is also observed that the results of the original MUSIC approach become rough around the other error peak estimations in Figure 6c. These results clearly indicate that accuracy became lower in Figure 6c due to the decrease of the SNR value. In contrast, for the results in Figure 6d–f, the estimation approach showed clear and accurate results when the VAD-MUSIC algorithm was used. There are no error peak estimations in the various peaks, which indicates that the VAD-MUSIC algorithm can account for the interruption of stationary sound that is similar to white noise.

Table 1. The settings of CNC milling operation.

	Spindle Speed (rpm)	Depth of Cut (mm)	Feed Rate (mm/min)
Data1	4000	1	600
Data2	4000	5	600
Data3	4000	5	500
Data4	8000	5	500

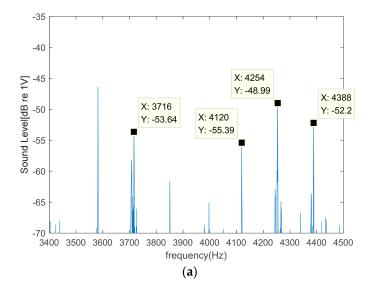


Figure 4. Cont.

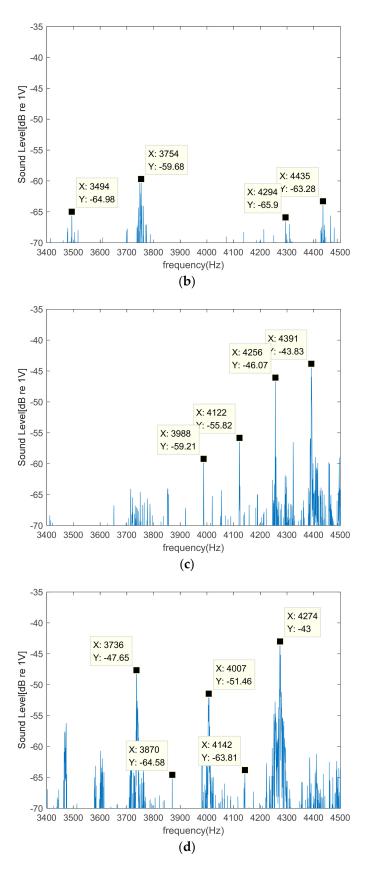


Figure 4. The CNC milling chatter circumstances are detected at a high frequency between 3.4–4.5 kHz (a) resonance frequencies of data1; (b) resonance frequencies of data2; (c) resonance frequencies of data4.

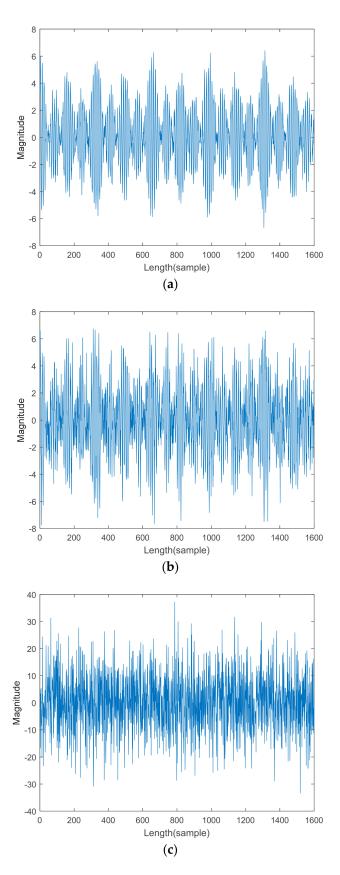


Figure 5. The input of time domain multi-tone signals with different SNRs. (a) SNR = 12 dB; (b) SNR = 6 dB; (c) SNR = 0 dB.

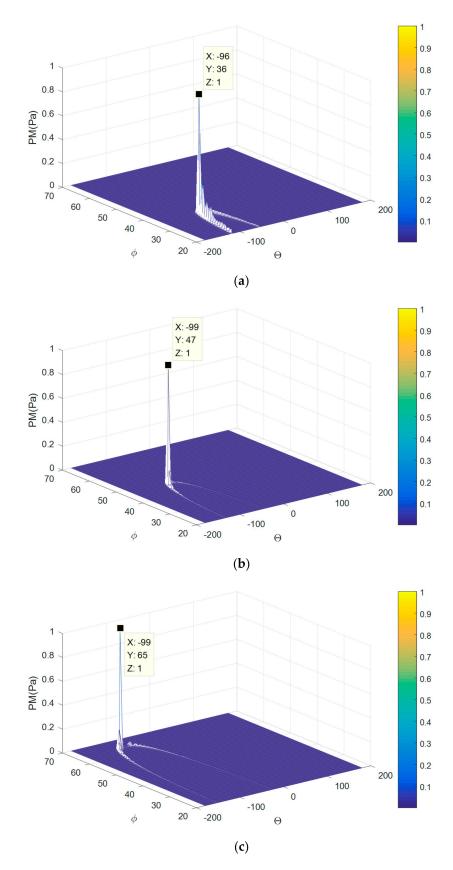


Figure 6. Cont.

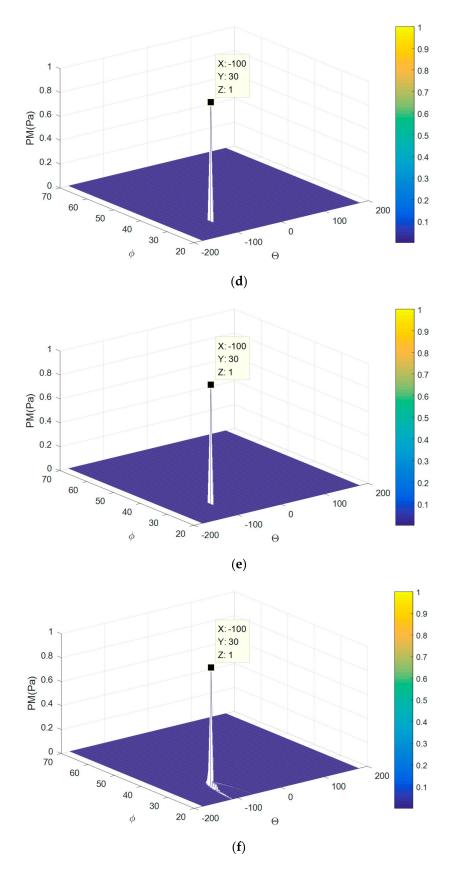


Figure 6. Source $(\theta, \varphi) = (-100, 30)$ estimation in high noise environments with **(a)** SNR = 12 dB, **(b)** SNR = 6 dB, and **(c)** SNR = 0 dB for the original MUSIC method and **(d)** SNR = 12 dB, **(e)** SNR = 6 dB, and **(f)** SNR = 0 dB for the VAD + MUSIC method.

4. Conclusions

In this study, an algorithm utilizing both the VAD and MUSIC methods for the noise location estimation of CNC milling chatter is proposed. The results in different SNR cases show that the proposed method is highly accurate at detecting CNC milling chatter locations when the SNR is high. In the future, we are going to incorporate the VAD-MUSIC algorithm into embedded microphone array systems to enable the practical detection of the sources of CNC milling chatter during real time processing.

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Conflicts of Interest: The authors declare no conflict of interest.

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